Analyzing Circuits with Impedances

Ohm's Law:

$$V = IZ \tag{1}$$

Where:¹

$$Z_R = R \tag{2}$$

$$Z_L = i\omega L \tag{3}$$

$$Z_C = 1/i\omega C \tag{4}$$

In series:

$$Z = Z_1 + Z_2 + \dots + Z_n \tag{5}$$

In parallel:

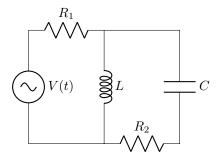
$$\frac{1}{Z} = \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_n}$$
(6)

Example: Amplitudes when driven by voltage

When we are only interested in voltage and current amplitudes and not phases, we use:

$$Z \quad \mapsto \quad |Z| = \sqrt{[\operatorname{Re}(Z)]^2 + [\operatorname{Im}(Z)]^2} \tag{7}$$

Consider the circuit below with $V(t) = V_0 \cos(\omega t)$:



What is the impedance of everything to the right of the battery? Express in the form $Z = R + i\chi$ for real R and χ .

Using series and parallel rules followed by complex arithmetic:²

$$Z = Z_{R_1} + \frac{1}{\frac{1}{Z_L} + \frac{1}{Z_C + Z_{R_2}}}$$
(8)

$$= Z_{R_1} + \frac{Z_L(Z_C + Z_{R_2})}{Z_C + Z_{R_2} + Z_L}$$
(9)

$$= R_1 + \frac{i\omega L((1/i\omega C) + R_2)}{(1/i\omega C) + R_2 + i\omega L}$$
(10)

$$= R_1 + \frac{L/C + i(\omega L R_2)}{R_2 + i(\omega L - 1/\omega C)} \cdot \frac{R_2 - i(\omega L - 1/\omega C)}{R_2 - i(\omega L - 1/\omega C)}$$
(11)

$$= R_1 + \frac{[LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)] + i[(\omega LR_2^2) - (L/C)(\omega L - 1/\omega C)]}{R_2^2 + (\omega L - 1/\omega C)^2}$$
(12)

$$= \left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2} \right] + i \left[\frac{(\omega LR_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2} \right]$$
(13)

 $^1Z=\overline{{\rm Re}(Z)+i\,{\rm Im}(Z)=R+i\chi}$ for resistance R and reactance $\chi.$ $^2i^2=-1$ so 1/i=-i

What is the maximum current amplitude through R_1 ?

We consider Ohm's law as $V^{\max} = I^{\max}|Z|$, with Z given by (13), with the current at its maximum amplitude when the voltage is at its maximum amplitude $V(t) = V_0$. Then the current through R_1 is the current that goes through the circuit as a whole and it given by:

$$I_{R_1}^{\max} = \frac{V_0}{|Z|} \tag{14}$$

$$= \frac{v_0}{\sqrt{\left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2 + \left[\frac{(\omega LR_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2}}$$
(15)

What is the maximum voltage amplitude through R_2 ?

By charge conservation, we know that the current that flows through the inductor plus the current that flows through R_2 is equal to the current that flows through R_1 , so we have:

$$I_{R_1} = I_L + I_{R_2} \implies I_{R_1}^{\max} = I_L^{\max} + I_{R_2}^{\max}$$
(16)

Additionally, since energy is single-valued, the voltage drop across the inductor must equal the voltage drop across the capacitor and R_2 , so:

$$I_L Z_L = I_{R_2} (Z_C + Z_{R_2}) \tag{17}$$

Now, working with magnitudes, $I_{R_1} = I_{R_1}^{\max}$ as in (14-15), and $|Z_L| = \omega L$, and $|Z_C + Z_{R_2}| = \sqrt{R_2^2 + 1/(\omega C)^2}$, so we have two equations and two unknowns, solving we find:

$$I_L^{\max} = \frac{|Z_C + Z_{R_2}|^2}{|Z_L| + |Z_C + Z_{R_2}|} I_{R_1}^{\max}$$
(18)

$$I_{R_2}^{\max} = \frac{|Z_L| \cdot |Z_C + Z_{R_2}|}{|Z_L| + |Z_C + Z_{R_2}|} I_{R_1}^{\max}$$
(19)

Or, substituting in:

$$I_L^{\max} = \frac{R_2^2 + 1/(\omega C)^2}{\omega L + \sqrt{R_2^2 + 1/(\omega C)^2}} \cdot \frac{V_0}{\sqrt{\left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2 + \left[\frac{(\omega LR_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2}} \qquad (20)$$

$$I_L^{\max} = \frac{\omega L\sqrt{R_2^2 + 1/(\omega C)^2}}{V_0} \cdot \frac{V_0}{V_0} \qquad (21)$$

$$I_{R_2}^{\max} = \frac{\sqrt{2} + (L/2)}{\omega L + \sqrt{R_2^2 + 1/(\omega C)^2}} \cdot \frac{1}{\sqrt{\left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2 + \left[\frac{(\omega LR_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]^2}}$$
(21)

Thus, the maximum voltage amplitude through R_2 is:

$$V_{R_2}^{\max} = I_{R_2}^{\max} R_2 \tag{22}$$

with $I_{R_2}^{\max}$ as in (21).

What is the phase difference between the voltage in R_1 and R_2 ?

Since both components are resistors, this is equivalent to asking the phase difference between the currents. Now, we cannot use $I_{R_1}^{\max}$ and $I_{R_2}^{\max}$ since these are magnitudes with no phase information.

We have:

$$I_{R_1}(t) = \frac{V_0 \cos(\omega t)}{Z} \tag{23}$$

$$= \frac{V_0 \cos(\omega t)}{\left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}}\right]^2}$$
(24)

$$\times \left(\left[R_1 + \frac{LR_2/C + (\omega L - 1/\omega C)^2}{R_2^2 + (\omega L - 1/\omega C)^2} \right] - i \left[\frac{(\omega L R_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2} \right] \right)$$
(25)

In general, the phase is:

$$\theta(f(t)) = \tan^{-1} \left(\frac{\operatorname{Im}(f(t))}{\operatorname{Re}(f(t))} \right)$$
(26)

So, we see:

$$\theta_{R_1} = \tan^{-1} \left(-\frac{\frac{(\omega L R_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}}{R_1 + \frac{L R_2/C + (\omega L R_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}} \right)$$
(27)

$$= \tan^{-1} \left(-\frac{(\omega L R_2^2) - (L/C)(\omega L - 1/\omega C)}{R_1 (R_2^2 + (\omega L - 1/\omega C)^2) + L R_2 / C + (\omega L R_2)(\omega L - 1/\omega C)} \right)$$
(28)

Now:

$$I_{R_2}(t) = \frac{Z_L(Z_C + Z_{R_2})}{Z_L + Z_C + Z_{R_2}} I_{R_1}(t)$$
(29)

$$=\frac{[LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)] + i[\omega LR_2^2 - (L/C)(\omega L - 1/\omega C)]}{R_2^2 + (\omega L - 1/\omega C)^2}I_{R_1}(t)$$
(30)

$$=\frac{[LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)] + i[\omega LR_2^2 - (L/C)(\omega L - 1/\omega C)]}{1}$$
(31)

$$\times \frac{1}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \frac{V_{0} \cos(\omega t)}{\left[\sum_{\alpha} LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C)\right]^{2} - \left[(\omega LR_{2}^{2}) - (L/C)(\omega L - 1/\omega C)\right]^{2}}$$
(32)

$$\frac{R_{2} + (\omega L - 1/\omega C)^{2}}{R_{2} + (\omega L - 1/\omega C)^{2}} \left[R_{1} + \frac{LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right] + \left[\frac{(\omega LR_{2}) - (L/C)(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right] \times \left(\left[R_{1} + \frac{LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right] - i \left[\frac{(\omega LR_{2}) - (L/C)(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right] \right)$$
(33)

$$=\frac{1}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \int \frac{V_{0} \cos(\omega t)}{[V_{0} \cos(\omega t)]^{2}} \int \frac{V_{0} \cos(\omega t)}{[V_{0} \cos(\omega t)]^{2}}$$
(34)

$$= R_{2}^{2} + (\omega L - 1/\omega C)^{2} \left[R_{1} + \frac{LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right]^{2} + \left[\frac{(\omega LR_{2}^{2}) - (L/C)(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}} \right]^{2}$$

$$\left\{ \left(LR_{2} + (\omega LR_{2})(\omega L - 1/\omega C) \right) \right\} = LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C) \right\}$$
(61)

$$\times \left\{ \left(\left[LR_2/C + (\omega LR_2)(\omega L - 1/\omega C) \right] \left[R_1 + \frac{LR_2/C + (\omega LR_2)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2} \right] \right\}$$
(35)

$$+ \left[\omega L R_2^2 - (L/C)(\omega L - 1/\omega C)\right] \left[\frac{(\omega L R_2^2) - (L/C)(\omega L - 1/\omega C)}{R_2^2 + (\omega L - 1/\omega C)^2}\right]$$
(36)

$$+i\left(\left[\omega LR_{2}^{2}-(L/C)(\omega L-1/\omega C)\right]\left[R_{1}+\frac{LR_{2}/C+(\omega LR_{2})(\omega L-1/\omega C)}{R_{2}^{2}+(\omega L-1/\omega C)^{2}}\right]$$
(37)

$$-\left[LR_{2}/C + (\omega LR_{2})(\omega L - 1/\omega C)\right] \left[\frac{(\omega LR_{2}^{2}) - (L/C)(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right]\right)$$
(38)

So:

$$\theta_{R_{2}} = \tan^{-1} \left(\frac{\left[\omega L R_{2}^{2} - (L/C)(\omega L - 1/\omega C)\right] \left[R_{1} + \frac{L R_{2}/C + (\omega L R_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right] - \left[L R_{2}/C + (\omega L R_{2})(\omega L - 1/\omega C)\right] \left[\frac{(\omega L R_{2}^{2}) - (L/C)(\omega L - 1/\omega C)^{2}}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right]}{\left[L R_{2}/C + (\omega L R_{2})(\omega L - 1/\omega C)\right] \left[R_{1} + \frac{L R_{2}/C + (\omega L R_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right] + \left[\omega L R_{2}^{2} - (L/C)(\omega L - 1/\omega C)\right] \left[\frac{(\omega L R_{2}^{2}) - (L/C)(\omega L - 1/\omega C)^{2}}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right]}{\left(R_{1} + \frac{L R_{2}/C + (\omega L R_{2})(\omega L - 1/\omega C)}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right] + \left[\omega L R_{2}^{2} - (L/C)(\omega L - 1/\omega C)\right] \left[\frac{(\omega L R_{2}^{2}) - (L/C)(\omega L - 1/\omega C)^{2}}{R_{2}^{2} + (\omega L - 1/\omega C)^{2}}\right]}\right] \right)$$

$$(39)$$

If we have a 1 kHz signal with both resistors of 10 Ω , a capacitor of 1000 μ F, and a 1 mH inductor, then:

$$\theta_{R_1} = -19.37^{\circ} \tag{40}$$

$$\theta_{R_2} = 38.24^{\circ} \tag{41}$$

where V_{R_1} lags the voltage by 19.37°, and V_{R_2} leads the voltage by 38.24°, for a phase difference of:

$$|\theta_{R_1} - \theta_{R_2}| = 57.61^{\circ} \tag{42}$$